Passive Radar - From Inception to Maturity

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Outline

- Introduction and definitions
- Some history
- Bistatic radar properties: geometry, radar equation, target properties
- Passive radar illuminators
- Passive radar systems and results
- The future …
BISTATIC RADAR: DEFINITIONS

- **MONOSTATIC RADAR**
  Tx & Rx at same, or *nearly the same*, location

- **BISTATIC RADAR**
  Tx & Rx separated by a considerable *distance* in order to achieve a technical, operational or cost benefit

- **RADAR NET**
  Several radars linked together to improve *coverage* or accuracy

- **MULTILATERATION RADAR**
  Radar net using range-only data

- **MULTISTATIC RADAR**
  Bistatic radar net with multiple Txs and/or RXs.

- **HITCHHIKER**
  Bistatic Rx operating with the Tx of a monostatic radar

- **PASSIVE BISTATIC RADAR**
  Bistatic Rx operating with other Txs of opportunity

* *Enjoys the union of individual coverage areas. All others require the intersection of individual coverage areas.*

Coverage area: \((\text{SNR} + \text{BW} + \text{LOS})\)
BISTATIC RADAR

• Bistatic radar has potential advantages in detection of stealthy targets which are shaped to scatter energy in directions away from the monostatic

• The receiver is covert and therefore safer in many situations

• Countermeasures are difficult to deploy against bistatic radar

• Increasing use of systems based on unmanned air vehicles (UAVs) makes bistatic systems attractive

• Many of the synchronisation and geolocation problems that were previously very difficult are now readily soluble using GPS, and

• The extra degrees of freedom may make it easier to extract information from bistatic clutter for remote sensing applications
Bistatic Radar

- The first radars were bistatic (till T/R switches were invented)

- First resurgence (1950 – 1960): semi-active homing missiles, SPASUR, …

- Second resurgence (1975 – 1985): SANCTUARY, hitchhikers, multistatic measurement system (Kwajalein), …

THE FIRST RADAR — HULSMUEYER, 1904

No 13,170  A.D. 1904

Date of Application, 30th June, 1903—Accepted, 22nd Sept., 1904

COMPLETE SPECIFICATION.

"Herzian-wave Projecting and Receiving Apparatus Adapted to Indicate or Give Warning of the Presence of a Metallic Body, such as a Ship or a Train, in the Line of Projection of such Waves."—

1. CHRISTIAN HULSMUEYER of Gumbelstrasse, Dusseldorf, Germany, Engineer.

This invention consists broadly of improved apparatus for projecting electric waves in any desired direction combined with improved apparatus for receiving said waves when reflected back from any metallic body, such as a ship or a train, and receiving apparatus being adapted to put into action an audible or visible signal and thus give warning of the presence of such metallic body in the line of projection of the waves.

My invention is based upon the property of electric waves of being reflected back towards their source on meeting a metallic body, and will be more clearly understood by imagining a transmitting and receiving station such as indicated placed side by side at the same point and so arranged that waves projected from the transmitter can only strike the receiver by being reflected from some metallic body, which, at sea, would presumably be another ship.

I have illustrated my invention in the accompanying drawings, in which:

Fig. 1 is a longitudinal section showing a ship A fitted with my apparatus, and a ship B whose presence is detected thereby.

Fig. 2 is a sectional view of the apparatus, and

Figs. 3 and 4 are sectional views of details thereof.

My apparatus comprises a transmitting and a receiving station similar to those used in wireless telegraphy, with this difference that the two stations are situated in close proximity to each other and are so arranged and connected that they can be directly influenced one another. In view of the fact that ships are made subject to considerable rolling, pitching and heaving motion, which might otherwise render the apparatus practically useless, I provide both the transmitting and receiving antennas to a compass so that in Figs. 2, 3 and 4, which are seen in a sectional view, I have illustrated a suitable transformer. The secondary current of the induction coil is conducted by wire through a hollow cylinder to two insulated rings, fixed therein. On said spindle e I rotateably mount a sleeve d on the axis of a second spiral coil, e, which is rigidly held with a bar-shaped collector or frame, f, adapted to commutate the electric waves coming from the coil c and to assist in supporting them in their desired direction. The high tension, and correspondingly insulated current, from the induction coil is taken off the rings, by means of brushes, g, and transmitted to the condenser, h, in position which is mounted whereby information is gathered by the condenser and the receiving rings, i, on the receiving station.
THE FIRST RADAR – HULSMEYER, 1904
**First Passive Bistatic Radar**

The BBC Empire transmitter at Daventry gave a beam $30^\circ$ azimuth $\times 10^\circ$ elevation at 49 m wavelength, and the beat note from a Heyford bomber at a range of 8 miles was clearly detected.
THE DAVENTRY EXPERIMENT: 26 FEBRUARY 1935
THE DAVENTRY EXPERIMENT: 26 FEBRUARY 1935

From Watson Watt: ‘The hum of the expected Heyford from R.A.E. became audible, and we watched him fly by at about 6,000 feet towards Daventry. His instructions were to shuttle at that height to-and-fro on a twenty-mile long beat from Daventry, on a course up and down the centre-line of the radio beam. …. He made only a fair job of holding the requested course, no one of his four runs took him right over our heads, but three passed very close.’

From Wilkins: ‘The second approach was nearer the beam axis but still some way off and this time rhythmic beating of the re-radiated signal with a small direct signal allowed through the receiver was noted. As the aircraft subsequently flew off to the south good beats were observed and, calculating from the time interval from the airspeed requested (100 mph), we estimated that we had followed the aircraft for about eight miles'.
KLEIN HEIDELBERG

KLEIN HEIDELBERG
There were six KH *Stellungen* (sites):

- **BIBER** (Oostvoorne)
- **BREMSE** (Ostend)
- **BULLDOGGE** (Boulogne)
- **SKORPION** (Vaudricourt, Abbeville)
- **AUERHAHN** (Cap d’Antifer)
- **TAUSENDFÜSSLER** (Cherbourg)

Some sources also include Castricum (Netherlands) and Rømø (Denmark), but these were ELEFANT and SEE-ELEFANT/RÜSSEL radars, whose receive antennas were similar to those of KH.

- 1st order *Stellungen*
- 2nd order *Stellungen*
This Report is not to be shown outside the following distribution without the authority of A.C.A.S.(I).

**WAR CABINET**

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<th>Name</th>
<th>Number</th>
</tr>
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<td>Lord Cherwell</td>
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<td>Deputy A.C.O. of S., A.E.</td>
<td>26</td>
</tr>
<tr>
<td>C.S.O., Bomber Command.</td>
<td>27</td>
</tr>
<tr>
<td>C.I.O., Bomber Command.</td>
<td>28</td>
</tr>
<tr>
<td>C.S.O., Fighter Command.</td>
<td>29</td>
</tr>
<tr>
<td>C.I.O., Fighter Command.</td>
<td>30</td>
</tr>
<tr>
<td>Commanding General.</td>
<td>31</td>
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<td>U.S.S.T.A.F.</td>
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<td>Commanding General.</td>
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<td>A.O.C., 100 Group.</td>
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<td>A.O.C., 60 Group.</td>
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<td>37</td>
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<td>39</td>
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<td>D.H.</td>
<td>40</td>
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<tr>
<td>D.B.</td>
<td>41</td>
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<tr>
<td>D.O.</td>
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<tr>
<td>D.O.</td>
<td>43</td>
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<tr>
<td>H.A.P.</td>
<td>44</td>
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**D.C.D.**

<table>
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<tr>
<th>Name</th>
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<tbody>
<tr>
<td>Chief Superintendent, T.R.E.</td>
<td>35</td>
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</tbody>
</table>

**INTRODUCTION**

1.1. The enemy has been attempting to use the radiations from our own CH stations to plot our aircraft through the Mandrel screen. The evidence from the various sources is reviewed and possible countermeasures are considered.

**INTELLIGENCE**

2.1. A.D.I(K) has provided the only direct evidence; this comes from the interrogation of a prisoner, previously a Wassermann operator, who was posted last year to the Wassermann site at Abbeville / Vaucouleurs. According to his story the apparatus was equipped with a receiver, but no transmitter. Plots were obtained by locking on to the pulses from the Dover CH station. The maximum range claimed each day was usually about 450 km; the range accuracy was said to be ± 10 km.

2.2. The method is simple. The reading on the range tube gives the difference between (i) the distance from the CH to the Wassermann and (ii) the length of the path CH - aircraft - Wassermann. Since (i) is fixed and known, this determines (ii). Hence the aircraft must lie on an ellipse whose focus are the CH and the Wassermann. The position of the aircraft on the ellipse is then determined by taking a bearing. The prisoner declared that the Wassermann also gave the height of the aircraft.

2.3. The prisoner referred to the presentation unit of the Wassermann as 'Heidelberg'.

2.4. The Wassermann at Vaucouleurs is known from photographic evidence to have a wider array than normal. This has only been seen on seven apparatuses, some of which carry the normal array as well. They are listed below:

- Limen / Dulmen: 52° 35' 02" N 04° 37' 22" E
- Ostvoorne / Zechendon: 51° 54' 16" N 04° 03' 30" E
- Osten / de Nahn: 51° 10' 49" N 03° 02' 56" E
- Boulogne Monument: 50° 44' 46" N 01° 36' 52" E
- Abbeville / Vaucouleurs: 50° 07' 30" N 01° 32' 00" E
- Cap d'Antifer: 49° 40' 34" N 00° 09' 45" E
- Cherbourg / La Brasserie: 49° 40' 05" N 01° 28' 10" E

The Boulogne Wassermann had the large array by about August 1943. The remainder were not completed until towards the end of the year.

2.5. The German Air Signals Experimental Regiment is known to have a detachment at Ostvoorne. They refer in their correspondence to an apparatus called Klein Heidelberg. Since we know that there is a Wassermann at Ostvoorne similar to that at Vaucouleurs, we may reasonably suppose that they are for the same purpose. The relation between Heidelberg and Klein Heidelberg is not known, but it is assumed to be close.

2.6. A further detachment of the Air Signals Experimental Regiment is known to have arrived at Ostvoorne in December 1943.
from Boulogne. Ostend was not operational at that time but it was expected to be so shortly. It looks therefore, as if the Boulogne site was operational during the later part of 1942 and that experienced personnel from there had been sent to start up the Ostend site.

2.7. We have no further information concerning the remaining sites.

2.8. Captured documents dated November 1942 have referred to "Heidelberg" being under development.

VALUE TO THE ENEMY

3.1. From the above evidence it seems clear that the enemy has used the method, if only on a small scale. During the first half of this year he was getting from his radar, and latterly from his route-tracing organisations, an abundance of information which would have observed any additional long range plots obtained by the Heidelberg method. The present situation is something of a mystery however. For the last few months his main trouble in deploying his night fighters has been lack of early warning owing to the Mandrel Screen and to radio silence on the part of the bomber force.

3.2. It is inconceivable that the enemy would neglect any available method of obtaining early warning. Hence he must have encountered some difficulty, of whose nature we are unaware, in using this system, at least on the Dutch stations. While the present picture is not clear, the method represents a threat to our bomber offensive which we cannot neglect.

OTHER SUITABLE TRANSMISSIONS

4.1. In order to be fairly sure of getting a plot over any desired area by this method it is only necessary that the transmitter concerned be of reasonably high power and wide angle, and emit suitably sharp pulses. Apart from the CH's themselves, the CH stand-by sets (45-50 Mc/s) and the Gee transmitters are the most suitable. Although there is as yet no evidence that the enemy is using the latter transmissions, we should bear the possibility in mind.

POSSIBLE COUNTERMEASURES

5.1. Provided that we can be reasonably sure of identifying the responsible apparatus a simple countermeasure would be to destroy them. The main difficulty is that, while we can be reasonably sure of identifying Chimeras on the coast, we cannot be sure of finding them inland, nor of finding other types of apparatus which might be used.

5.2. It has been proposed as a countermeasure to put two CH stations on the same frequency and "jitter" their pulses. This would not be satisfactory because (i) the enemy could employ a directional aerial, and so select whichever station he preferred, and (ii) even if he did not do this but merely triggered his time base at every direct pulse, he would only develop an ambiguity, which could probably be resolved by a few minutes' plotting, and which would in any case give him the early warning he so badly wants.

5.3. With the CH stations and their stand-by station and better radio-countermeasures suggest themselves. The first and obvious one is to switch them off as the Mandrel Screen comes on. A second and rather nice one is to fit a "Moonshine" modification so as to provide the enemy with false plots. It would be possible in this way to decide whether or not the enemy is at present using the method; we could easily simulate a force of bombers flying across the North Sea and observe the enemy's reactions. If successful the device could be used to provide so many false alarms that he could no longer trust the observations.

5.4. Should he decide to use Gee transmissions, radio-countermeasures might be difficult without interfering with our own use of this aid. In this case destruction of the receiving apparatuses responsible may be the best alternative.

24.11.44. A.D.I. (Science).
KLEIN HEIDELBERG

L480 Bunker (Stellung BIBER, Oostvoorne)
**Bistatic Radar Geometry**

\[ R_R = \frac{(R_T + R_R)^2 - L^2}{2(R_T + R_R + L \sin \theta_R)} \]

\[ \beta = \theta_T - \theta_R \]

Bistatic Radar Geometry

Contours of constant bistatic range are ellipses, with the transmitter and receiver as the two focal points.

\[ R_T + R_R = \text{const} \]

Targets lying on the transmitter-receiver baseline have zero bistatic range.
**Bistatic Radar Doppler**

\[ \beta = \theta_T - \theta_R \]

BISTATIC RADAR DOPPLER

For $V_T = V_R = 0$ ; $V \neq 0$ \[ f_D = \left( \frac{2V}{\lambda} \right) \cos \delta \cos(\beta/2) \]

special cases :

<table>
<thead>
<tr>
<th>$\beta$</th>
<th>$\delta$</th>
<th>$f_D$</th>
<th>condition</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>–</td>
<td>$(2v/\lambda) \cos \delta$</td>
<td>monostatic</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>$(2v/\lambda)$</td>
<td>monostatic</td>
</tr>
<tr>
<td>180</td>
<td>–</td>
<td>0</td>
<td>forward scatter</td>
</tr>
<tr>
<td>–</td>
<td>$\pm 90$</td>
<td>0</td>
<td>$v \perp$ to bisector</td>
</tr>
<tr>
<td>–</td>
<td>$\pm \beta/2$</td>
<td>$(2v/\lambda) \cos^2(\beta/2)$</td>
<td>$v \Rightarrow tx$ or $rx$</td>
</tr>
<tr>
<td>–</td>
<td>0,180</td>
<td>$\pm (2v/\lambda) \cos(\beta/2)$</td>
<td>$v \Rightarrow$ bisector</td>
</tr>
<tr>
<td>–</td>
<td>$90 \pm \beta/2$</td>
<td>$\mp (v/\lambda) \sin \beta$</td>
<td>$v \perp$ to tx or rx LOS</td>
</tr>
</tbody>
</table>

This is derived in the same way as the monostatic radar equation:

\[
\frac{P_R}{P_N} = \frac{P_T G_T}{4\pi R_T^2} \cdot \sigma_b \cdot \frac{1}{4\pi R_R^2} \cdot \frac{G_R \lambda^2}{4\pi} \cdot \frac{1}{kT_0 BFL} = \frac{P_T G_T G_R \lambda^2 \sigma_b}{(4\pi)^3 R_T^2 R_R^2 kT_0 BFL}
\]

The dynamic range of signals to be handled is reduced, because of the defined minimum range.
We can see from the bistatic radar equation

$$\frac{P_R}{P_N} = \frac{P_T G_T G_R \lambda^2 \sigma_b}{(4\pi)^3 R_T^2 R_R^2 kT_0 BFL}$$

that contours of constant detection range are defined by $R_T R_R = \text{constant} = c$.

These are *Ovals of Cassini*


**Resonance Scatter**

- The resonance scatter effect for monostatic radars has been well documented for conventional (non-stealth) targets. In the simplest case of a conducting sphere of radius \(a\), resonance occurs in the region \(0.5 < \frac{2\pi a}{\lambda} < 10\).

- "Physically, the resonant region can be explained by the interference between the incident wave and the creeping wave, which circles the sphere and either adds to or subtracts from the total field at the leading surface" (Barton)

- The net result of these additive effects is that when wavelengths are of the order of discrete aircraft dimensions, for example fuselage, wing, tail, inlet and exhaust ducts, the resulting resonance significantly enhances RCS when compared to the optical region, which for the sphere starts at \(\frac{2\pi a}{\lambda} > 10\).
Monostatic radar cross section of a simple, wire-grid aircraft model using method-of-moments computed patterns, TE-polarized incident plane wave. The peak in the curve around 270 MHz is due to a resonance condition for the fuselage. Resonances occur at several frequencies due to various aircraft parts and are most pronounced for slender metallic shapes. The resonance effect for this model ends at about 400 MHz. Courtesy AIAA.

Deutsche Aerospace, Bremen anechoic chamber measurements of a faceted, metallized 1:10 scale model of an F-117 aircraft:

“The aircraft geometry was obtained from open literature and hence the target model does not take into account fine structure details and surface materials such as RAM, which are of less importance at VHF/UHF…”

“The scaled measurement results… show that the attempt to reduce the target’s RCS has been successful in the ±70° section around the nose-on aspect and for the frequency range above 400 MHz. High RCS values covering the whole frequency range occur when the direction of illumination is perpendicular to the front or back edges of the wings or other dominant structures of the fuselage. In the nose-on section, however, an increase in the RCS can be seen at VHF around 100 MHz [6 to 10 dBm²] and UHF around 400 MHz [0 to 6 dBm², > 6 dBm² nose on] due to resonance effects. Hence, such stealth techniques can be efficient at high radar frequencies but are ineffective at VHF/UHF.”

**FORWARD SCATTER**

Babinet’s principle tells us that we get exactly the same scattering from a perfectly-absorbing target as we would from a target-shaped hole in an infinite perfectly-conducting sheet!
**Resonance Scatter**

So a target on the transmitter-receiver baseline, even if it is completely stealthy, will scatter a significant amount of energy - in fact the RCS will be of the order of

$$\sigma_b = \frac{4\pi A^2}{\lambda^2}$$

The angular width of the scattering will be of the order of $\lambda/d$ (radians)

which tends to favour a low frequency.

But a target which lies exactly on the transmitter-receiver baseline will give no range information and no Doppler information, and even for a target only slightly off-baseline the range and Doppler resolution will be poor.

So whilst a forward scatter radar will be good for target detection, location and tracking will be more difficult.
BISTATIC/MONOSTATIC RCS FOR 4 SMALL FREIGHTERS

Assembly of discrete scattering centers
- Flat plates
- Corner reflectors
- Dihedrals
- Resonant cavities

PERFORMANCE PREDICTION

Maximum integration dwell time is approximately

\[ T_{MAX} = \left( \frac{\lambda}{A_R} \right)^{1/2} \]

For a VHF waveform with a bandwidth of 50 kHz and a dwell time of 1 second, processing gain is \( G_p \approx 47 \text{ dB} \).

Cast bistatic radar equation in the form:

\[ \left( R_R \right)_{max} = \left( \frac{\Phi \sigma_b G_R \lambda^2 G_p}{(4\pi)^2 (S/N)_{min} kT_0 BF} \right)^{1/2} \]

allows prediction of coverage around transmitters and receivers.
**VERTICAL PLANE COVERAGE**

- The performance of a passive bistatic radar system depends not only on the waveform, but also on the coverage of the illuminating sources.
- Transmitters will frequently be sited on hilltops or on tall buildings.
- The vertical-plane coverage will usually be optimized so as to avoid wasting power above the horizontal, and in some cases the beams may be tilted downwards.

\[
R_{R_{\text{max}}} = \sqrt{\frac{P_T G_t \lambda^2 \sigma G_p}{(4\pi)^3 R_T^2 (S/N)_{\text{min}} kT_0 BFL}}
\]

- For every 10 dB reduction in \( P_t G_t \) the maximum detection range \( R_R \) for a given target is reduced by a factor of 3.3 \( \times \)

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PASSIVE BISTATIC RADAR
PASSIVE BISTATIC RADAR

- Broadcast and communications transmitters tend to be sited on high locations and hence achieve broad coverage.
- Since the system makes use of existing transmitters, the cost of a passive radar is likely to be much lower than a conventional radar.
- Similarly, there are no licensing issues.
- It allows the use of frequency bands (particularly VHF and UHF) that are not normally available for radar purposes. Such frequencies may be beneficial in detecting stealthy targets, since the wavelength is of the same order as the physical dimensions of the target, and forward scatter gives a relatively broad angular scatter.
- Since the receiver emits no signal of its own, and as long as the receive antenna is inconspicuous, the passive radar receiver may be undetectable and hence completely covert.
- It is difficult to deploy countermeasures against passive radar. Any jamming will have to be spread over a range of directions, diluting its effectiveness.
- Passive radar does not require any additional spectrum. For this reason it has been termed ‘green radar’.
- There is an enormous range of transmissions that may be used. In practice, almost any emission can be used as the basis of a passive radar.
PASSIVE BISTATIC RADAR

- The waveforms of such transmissions are not optimized for radar purposes, so care has to be used to select the right waveforms and to process them in the optimum way.

- In many cases the transmit source is not under the control of the passive radar.

- For analog signals, the ambiguity function (resolution in range and in Doppler) depends on the instantaneous modulation, and some kinds of modulation are better than others. Digital modulation does not suffer from these problems, so is likely to be preferred.

- The waveforms are usually continuous (i.e. a duty cycle of 100%), so significant processing has to be used to suppress the direct signal and multipath in order to detect weak target echoes.

- In common with all bistatic radars, the resolution in range and Doppler is poor for targets on or close to the baseline between transmitter and receiver.
## Passive Radar Illuminators

<table>
<thead>
<tr>
<th>Transmission</th>
<th>Frequency</th>
<th>Modulation, Bandwidth</th>
<th>$P_iG_i$</th>
<th>Power density (note 1)</th>
</tr>
</thead>
<tbody>
<tr>
<td>HF broadcast</td>
<td>10–30 MHz</td>
<td>DSB AM, 9 kHz</td>
<td>50 MW</td>
<td>$-67$ to $-53$ dBW/m$^2$ at $R_T = 1,000$ km</td>
</tr>
<tr>
<td>VHF FM</td>
<td>88–108 MHz</td>
<td>FM, 200 kHz</td>
<td>250 kW</td>
<td>$-57$ dBW/m$^2$ at $R_T = 100$ km</td>
</tr>
<tr>
<td>Analog TV</td>
<td>~550 MHz</td>
<td>Vestigial sideband AM, 5.5 MHz</td>
<td>1 MW</td>
<td>$-51$ dBW/m$^2$ at $R_T = 100$ km</td>
</tr>
<tr>
<td>DAB</td>
<td>~220 MHz</td>
<td>Digital, OFDM, 220 kHz</td>
<td>10 kW</td>
<td>$-71$ dBW/m$^2$ at $R_T = 100$ km</td>
</tr>
<tr>
<td>DVB-T</td>
<td>~750 MHz</td>
<td>Digital, 6 MHz</td>
<td>8 kW</td>
<td>$-72$ dBW/m$^2$ at $R_T = 100$ km</td>
</tr>
<tr>
<td>Cell phone base station (GSM)</td>
<td>900 MHz, 1.8 GHz</td>
<td>GMSK, FDMA/TDMA/FDD, 200 kHz</td>
<td>100 W</td>
<td>$-71$ dBW/m$^2$ at $R_T = 10$ km</td>
</tr>
<tr>
<td>Cell phone base station (3G)</td>
<td>2 GHz</td>
<td>CDMA, 5 MHz</td>
<td>100W</td>
<td>$-41$ dBW/m$^2$ (note 2) at $R_T = 10$ m</td>
</tr>
<tr>
<td>WiFi 802.11</td>
<td>2.4 GHz</td>
<td>DSSS/OFDM, 5 MHz</td>
<td>100 mW</td>
<td>$-88$ dBW/m$^2$ at $R_T = 10$ km</td>
</tr>
<tr>
<td>WiMAX 802.16</td>
<td>2.4 GHz</td>
<td>QAM, 1.25–20 MHz</td>
<td>20W</td>
<td>$-134$ dBW/m$^2$ at Earth’s surface</td>
</tr>
<tr>
<td>GNSS</td>
<td>L-band</td>
<td>CDMA, FDMA, 1–10 MHz</td>
<td>200W</td>
<td>$-107$ dBW/m$^2$ at Earth’s surface</td>
</tr>
<tr>
<td>DBS TV</td>
<td>Ku-band, 11–12 GHz</td>
<td>Analog and digital</td>
<td>300 kW</td>
<td>$-54$ dBW/m$^2$ SAR (note 3) at Earth’s surface</td>
</tr>
<tr>
<td>Satellite SAR</td>
<td>9.6 GHz</td>
<td>Chirp pulse, 400 MHz (max.)</td>
<td>28 MW</td>
<td></td>
</tr>
</tbody>
</table>

Source: [6]. Note 1: assuming free space line-of-sight propagation; Note 2: would be subject to additional attenuation due to propagation through walls; and Note 3: parameters from SARs carried by the COSMO-SkyMed series of satellites [7].
Locations of CONUS FM Stations

- 4500 FM Transmitters ≥ 5 Kwatts
- 3600 FM Transmitters ≥ 1 Kwatt

High Power FM Line-of-Sight for 1000’ Targets

Existing commercial FM transmitters provide low altitude illumination for virtually all regions of interest in the US.

For Europe, CIA World Factbook cites:
- >15000 FM Stations
- >44000 TV Stations
PASSIVE RADAR WAVEFORMS

- BBC Radio 4 (news) 93.5 MHz
- Jazz FM (fast tempo) 102.2 MHz
- Digital Audio Broadcast 222.4 MHz
"Entirely possible" is the scientific verdict of radio engineers at the National Bureau of Standards to British dispatches citing the use of television receivers as "spotters" of airplanes.

While Army officials would not confirm reports that similar methods are being worked out for the military uses of the United States, it was admitted that secret research is underway to test other ways of spotting airplanes than by the present sound detection methods.

Since television broadcasts have been in progress over London it has been noted that when airplanes are flying in the vicinity there are produced "ghost" images in the television receivers. These "ghosts" are caused by reflection of the television waves from the metal airplane surface. Thus the reflected waves arrive at the television receivers at a slightly different time than the ordinary waves. The result is a dual image of the scene being transmitted. The image of the plane itself is not received.

According to British reports the displacement of the "ghost" image has been correlated with the distance of the plane away from the television receivers. A system has been worked out whereby television receivers on England's eastern coast could thus serve as "spotters" for approaching enemy aircraft in time of war.

Whether the plan can be worked out in complete detail and serve a valuable military use is for the future to decide, but in principle the method is an almost exact counterpart of the system of determining airplane altitude by having the plane send down to the ground a beam of radio waves and then having the plane pick up the signals of the reflected waves. This method was announced by Dr. E. F. W. Alexanderson of the General Electric Company in 1928.

For the television case, in contrast, the waves go up, strike the plane, and are picked up by ground receivers. By multiple receivers and methods of triangulation it is believed the altitude of the plane and its approximate direction and distance could be worked out.

In another analogy the television spotting system for planes can be called "upside-down" geophysical prospecting. In geology, metallic masses are located by reflected radio waves.
measured spectrum of analogue (and digital) TV signals

In the UK the PAL (Phase Alternating Line) modulation format is used, in which the video information is coded as two interlaced scans of a total of 625 lines at a frame rate of 50 Hz. The start of each line is marked with a sync pulse, and the total duration of each line is 64 μs. The video information is modulation onto a carrier as vestigial-sideband AM, coded as luminance (Red + Green + Blue) and two chrominance signals (Green – Blue) and (Red – Blue). The two chrominance subcarriers are in phase quadrature, so that they can be separately recovered. The sound information (including stereo information) is frequency-modulated onto a second carrier.
EARLY BISTATIC RADAR EXPERIMENTS AT UCL

illuminating radar (Heathrow)

3-PRF stagger

PRF 'flywheel' clock

scan 'flywheel' clock

video

PPI display
dipole antenna
EARLY BISTATIC RADAR EXPERIMENTS AT UCL


FORWARD SCATTER VHF FM

- BBC Radio 4, 93.5 MHz, Wrotham (Kent)
- Aircraft target on approach to Heathrow, crossing baseline
- Receiver located so that direct signal is weak, hence beat between direct signal and Doppler-shifted echo has maximum modulation
- Doppler shift goes through zero as target crosses baseline
Tracking of a Cessna 172 light aircraft using FM radio transmission

Courtesy of Craig Tong and Professor Mike Inggs, University of Cape Town
**Homeland Alerter HA-100**

- developed by THALES and ONERA in 2005

- FM Radio band 88 – 108 MHz

- 8 vertical dipoles

- Signal processing is composed of a space time algorithm to reject the direct path, estimation / correlation, regulation process and Detection / measurement. The data processing merges the detections from all 8 channels to tracks in Cartesian coordinates. The update rate is 1.5 s
The passive radar from Airbus Defence and Space demonstrated its ability to detect low-flying objects in mission scenarios similar to those of various European armed forces, among other locations, in mountainous terrain with areas of major radar shadow. At the same time, the system registered the latest airspace situation in a very short time, at ranges of up to 200 kilometres. A real-time networked system, involving two devices at different locations, also demonstrated that it can even be used in areas with a particularly restricted transmitter infrastructure.
MODERN PASSIVE RADAR

HA-100 Homeland Alerter - THALES (France)

GAMMA - FKIE (Germany)

AIRBUS Defence and Space (Europe)

SILENT GUARD - ERA (Czech Republic)

AULOS - SELEX (Italy)

DLW002 (China)

ALIM (Iran)
PBR with an Aircraft-Borne Receiver
PBR WITH AN AIRCRAFT-BORNE RECEIVER
"First and foremost, the absence of active radiation makes the system very stealthy, not allowing the enemy to detect and destroy it with anti-radar means; secondly, the system detects enemy attack and reconnaissance drones which observe radio silence mode; also, the system can effectively detect sophisticated enemy stealth drones of various types"
THE FUTURE …

- The spectrum problem and Commensal Radar
- Air Traffic Management (ATM)
- Indoor monitoring for Eldercare / Assisted Living
- Border / harbor surveillance
- Target recognition with Passive Radar
- Low cost Passive Radar
- The Intelligent Adaptive Radar Network
**Passive Radar: A $10 Billion opportunity**

Passive radar is different from traditional forms of radar in that it does not emit any electromagnetic radiation. Instead, it relies on reflections from other electromagnetic signals in the atmosphere in order to provide a radar picture. Passive radar provides a number of distinct advantages that will allow it to corner a significant portion of defense, homeland security, and civilian radar markets. In addition to cost-efficiency, passive radar is also covert, an effective counter to stealth technologies, and environmentally friendly.

The market for passive radar is still in its infancy, and few companies have developed effective, marketable systems. However, as the technology becomes more sophisticated and affordable, more and more competitors can be expected to enter the market, particularly in defense and homeland security.

**Related Research on ASDReports.com:**

**The Military and Civil Aviation Passive Radar Market: 2013 - 2023**

By the end of 2023, SNS Research expects passive radar technology investments to account for more than $10 billion in revenue, following a CAGR of nearly 36% between 2013 and 2023.

The ‘Military & Civil Aviation Passive Radar Market: 2013 – 2023’ report focuses on the two markets where passive radar technology has the greatest potential: civilian aviation and military radar applications.

The report presents vendor strategies, overall depictions of potential growth in both sectors, as well as detailed qualitative and quantitative analysis of global and regional drivers and limitations on market potential from 2013 till 2023.

The report comes with an associated excel datasheet covering quantitative data from all revenue projection forecasts presented within the report.
Griffiths, H.D. and Baker, C.J,  
*An Introduction to Passive Radar*

Developed by recognized experts in the field, this first-of-its-kind resource introduces the basic principles of passive radar technology and provides an overview of recent developments in this field and existing real passive radar systems. This book explains how passive radar works, how it differs from the active type, and demonstrates the benefits and drawbacks of this novel technology. Properties of illuminators, including ambiguity functions, digital vs. analog, digitally-coded waveforms, vertical-plane coverage, and satellite-borne and radar illuminators are explored.

Readers will find practical guidance on direct signal suppression, passive radar performance prediction, and detection and tracking. This resource provides concrete examples of systems and results, including analog TV, FM radio, cellphone base stations, DVB-T and DAB, HF skywave transmissions, indoor Wi-Fi, satellite-borne illuminators, and low-cost scientific remote sensing. Future developments and applications of passive radar are also presented.